ETR05054-001

### 36V Operation 0.6A Synchronous Step-Down DC/DC Converters

### **■**GENERAL DESCRIPTION

The XC9267 series are 36V operation synchronous step-down DC/DC converter ICs with a built-in P-channel MOS driver transistor and N-channel MOS switching transistor.

The XC9267 series has operating voltage range of  $3.0V \sim 36.0V$  and high-efficiency power supply up to an output current of 0.6A. Low ESR capacitors such as ceramic capacitors can be used for the load capacitor ( $C_L$ ).

A 0.75V reference voltage source is incorporated in the IC, and the output voltage can be set to a value from 1.0V to 25.0V using external resistors (R<sub>FB1</sub>, R<sub>FB2</sub>).

1.2MHz or 2.2MHz can be selected for the switching frequency.

The soft-start time is internally set to 2.0ms (TYP.), but can be adjusted to set a longer time using an external resistor and capacitor. With the built-in UVLO function, the driver transistor is forced OFF when input voltage becomes 2.7V or lower.

The output state can be monitored using the power good function.

Internal protection circuits include over current protection, short-circuit protection, and thermal shutdown circuits to enable safe use.

### APPLICATIONS

- Electric Meter
- Gas Detector
- Various Sensor
- Industrial Equipment
- Home appliance

### **■**FEATURES

Input Voltage Range : 3~36V(Absolute Max 40V)

Output Voltage Range 1.0~25V
FB Voltage : 0.75V±1.5%
Oscillation Frequency : 1.2MHz, 2.2MHz

Output Current : 0.6A

Control Methods : PWM control

Efficiency88%@12V→5V、300mA

Soft-start Time : Adjustable by RC

Protection Circuits : Over Current Protection
: Thermal Shutdown

Short-circuit Protection

Low ESR Ceramic Capacitor : Ceramic Capacitor

Packages : SOT-89-5 (Without Power Good)

USP-6C (With Power Good)

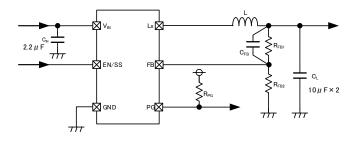
Environmentally Friendly EU RoHS Compliant, Pb Free

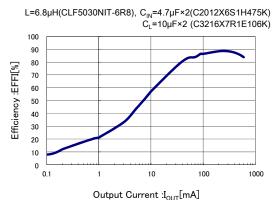
## ■TYPICAL APPLICATION CIRCUIT

# ■TYPICAL PERFORMANCE CHARACTERISTICS

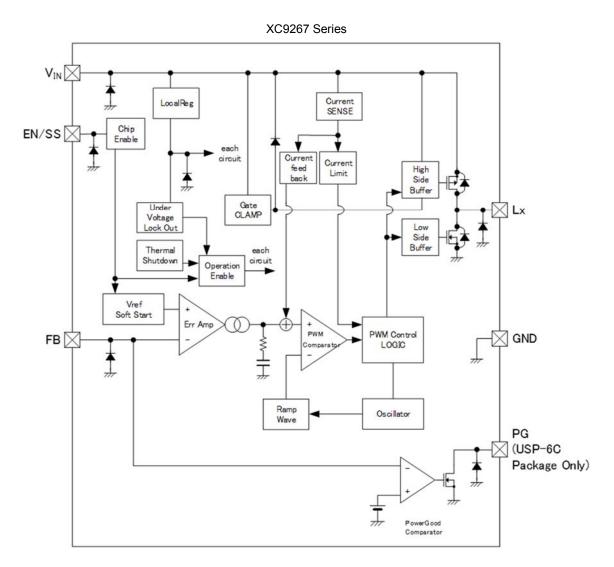
XC9267B75Cxx

( $V_{IN}$ =12V,  $V_{OUT}$ =5V,  $f_{OSC}$ =1.2MHz)





# **■BLOCK DIAGRAM**



Diodes inside the circuit are an ESD protection diodes and a parasitic diodes.

# **■PRODUCT CLASSIFICATION**

### Ordering Information

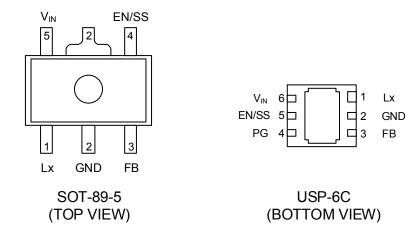
DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
1	Туре	В	Refer to Selection Guide
23	Adjustable Output Voltage	75	Output voltage can be adjusted in 1.0V to 25V
4	Oscillation Fraguency	С	1.2MHz
4	Oscillation Frequency	D	2.2MHz
56-7	Packages	PR-G <sup>(*1)</sup>	SOT-89-5 (1000pcs/Reel)
30-7	rackages	ER-G <sup>(*1)</sup>	USP-6C (3000pcs/Reel)

 $<sup>^{(*</sup>i)}$  The "-G" suffix denotes Halogen and Antimony free as well as being fully RoHS compliant.

### Selection Guide

TYPE	Packages	Chip Enable	UVLO	Thermal Shutdown	Power Good (USP-6C)	Soft Start	Current Limiter	Automatic Recovery (Current Limiter)
В	SOT-89-5	YES	YES	YES	NO	YES	YES	YES
В	USP-6C	YES	YES	YES	YES	YES	YES	YES

## **■PIN CONFIGURATION**



<sup>\*</sup> The dissipation pad for the USP-6C package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the GND (No. 2) pin.

## **■PIN ASSIGNMENT**

PIN NU	JMBER	PIN NAME	FUNCTION
SOT-89-5	USP-6C	FIIN INAIVIE	FUNCTION
5	6	Vin	Power Input
4	5	EN/SS	Enable Soft-start
-	4	PG	Power-good Output
3	3	FB	Output Voltage Sense
2	2	GND	Ground
1	1	Lx	Switching Output

## **■**FUNCTION CHART

PIN NAME	SIGNAL	STATUS	
	L	Stand-by	
EN/SS	EN/SS H Active		
	OPEN Undefined State(		

<sup>(1)</sup> Please do not leave the EN/SS pin open. Each should have a certain voltage.

# ■ ABSOLUTE MAXIMUM RATINGS

Ta=25°C

PARAN	PARAMETER		RATINGS	UNITS
V <sub>IN</sub> Pin	Voltage	Vin	-0.3 ∼ +40	V
EN/SS Pi	n Voltage	V <sub>EN/SS</sub>	-0.3 <b>∼</b> +40	V
FB Pin	Voltage	$V_{FB}$	-0.3 <b>~</b> +6.2	V
PG Pin V	′oltage <sup>(*1)</sup>	$V_{PG}$	-0.3 <b>~</b> +6.2	V
PG Pin (	Current <sup>(*1)</sup>	I <sub>PG</sub>	8	mA
Lx Pin \	Lx Pin Voltage		-0.3 ~ V <sub>IN</sub> + 0.3 or +40 (*2)	V
Lx Pin (	Current	I <sub>Lx</sub>	1800	mA
Power	SOT-89-5	Pd	1750(JEDEC board) (*4)	mW
Dissipation	USP-6C(DAF)	Fu	1250(JEDEC board) (*4)	IIIVV
Surge \	Surge Voltage		+46(*3)	V
Operating Ambie	Operating Ambient Temperature		-40 ~ +105	°C
Storage Te	emperature	Tstg	-55 <b>∼</b> +125	°C

<sup>\*</sup> All voltages are described based on the GND pin.

<sup>(\*1)</sup> For the USP-6C Package only.

 $<sup>^{(^{\</sup>ast}2)}$  The maximum value should be either  $V_{\text{IN}}\text{+}0.3$  or 40 in the lowest.

 $<sup>^{(*3)}</sup>$  Applied Time $\leq$ 400ms

<sup>(\*4)</sup> The power dissipation figure shown is PCB mounted and is for reference only. Please see the power dissipation page for the mounting condition.

## **■**ELECTRICAL CHARACTERISTICS

XC9267series Ta=25°C

PARAMETER	SYMBOL	CONE	DITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
FB Voltage	V <sub>FBE</sub>	V <sub>FB</sub> =0.739V→0.7 V <sub>FB</sub> Voltage wher changes from"H"	Lx pin voltage	0.739	0.750	0.761	V	2
Setting Output Voltage Range (*1)	Voutset	-		1	-	25	V	-
Operating Input Voltage Range (*1)	Vin	-		3	-	36	V	-
UVLO Detect Voltage	Vuvlod	V <sub>EN/SS</sub> =12V,V <sub>IN</sub> :2. V <sub>IN</sub> Voltage which holding "H" level	8V→2.6V,V <sub>FB</sub> =0V n Lx pin voltage	2.6	2.7	2.8	V	2
UVLO Release Voltage	Vuvlor	V <sub>EN/SS</sub> =12V,V <sub>IN</sub> :2. V <sub>IN</sub> Voltage which holding "L" level	7V→2.9V,V <sub>FB</sub> =0V n Lx pin voltage	2.7	2.8	2.9	V	2
0 : 10 1		.,	XC9267B75C	-	180	350		
Quiescent Current	Iq	V <sub>FB</sub> =0.825V	XC9267B75D	-	290	500	μA	4
Stand-by Current	I <sub>STBY</sub>	V <sub>IN</sub> =12V, V <sub>EN/SS</sub> =	V <sub>FB</sub> =0V	-	1.65	2.50	μA	4
Oscillation Frequency	fosc	Connected to external	XC9267B75C	1.098	1.200	1.302	MHz	MHz ①
Oscillation Frequency	IOSC	components, I <sub>OUT</sub> =200mA	XC9267B75D	2.013	2.200	2.387	1711 12	
Minimum On Time	tonmin	Connected to ext	ernal components	-	85 <sup>(*2)</sup>	-	ns	1
Minimum Duty Cycle	DMIN	V <sub>FB</sub> =0.825V		-	-	0	%	2
Maximum Duty Cycle	Dмах	V <sub>FB</sub> =0.675V		100	-	-	%	2
Lx SW "H" On Resistance	R <sub>LxH</sub>	V <sub>FB</sub> =0.675V, I <sub>Lx</sub> =2	200mA	-	1.20	1.38	Ω	(5)
Lx SW "L" On Resistance	R <sub>LxL</sub>			-	0.60	-	Ω	5
Highside Current Limit (*3)	I <sub>LIMH</sub>	V <sub>FB</sub> =V <sub>FBE</sub> ×0.98		1.00	1.30	-	Α	(5)
Internal Soft-Start Time	tss1	V <sub>FB</sub> =0.675V		1.6	2.0	2.4	ms	2
External Soft-Start Time	tss2	V <sub>FB</sub> =0.675V R <sub>SS</sub> =430KΩ, C <sub>SS</sub>	=0.47µF	21	26	33	ms	3
PG detect voltage (*4)	VPGDET	V <sub>FB</sub> =0.712V→0.638V, R <sub>PG</sub> :100kΩ pull-up to 5V V <sub>FB</sub> Voltage when PG pin voltage changes from"H" level to "L" level		0.638	0.675	0.712	V	(5)
PG Output voltage (*4)	$V_{PG}$	V <sub>FB</sub> =0.6V, I <sub>PG</sub> =1mA		-	-	0.3	V	2
Efficiency (*5)	EFFI	Connected to external components, V <sub>IN</sub> =12V, V <sub>OUT</sub> =5V, I <sub>OUT</sub> =300mA		-	88	-	%	1
FB Voltage Temperature Characteristics	ΔV <sub>FB</sub> / (ΔT <sub>opr</sub> •V <sub>FBE</sub> )	-40°C≦T <sub>opr</sub> ≦105	°C	-	±100	-	ppm/°C	2

Test Condition: Unless otherwise stated, V<sub>IN</sub>=12V, V<sub>EN/SS</sub>=12V, V<sub>PG</sub>:OPEN (\*4)

Peripheral parts connection conditions:

 $L\text{=}6.8\mu\text{H}, R_{\text{FB1}}\text{=}680\text{k}\Omega, R_{\text{FB2}}\text{=}120\text{k}\Omega, C_{\text{FB}}\text{=}18\text{pF}, C_{\text{L}}\text{=}10\mu\text{F} \times 2\text{parallel}, \ C_{\text{IN}}\text{=}2.2\mu\text{F}$ 

 $<sup>^{(\</sup>mbox{\tiny 1})}\mbox{Please}$  use within the range of  $\mbox{V}_{\mbox{\scriptsize OUT}}\mbox{/V}_{\mbox{\scriptsize IN}}\!\! \geq \! t_{\mbox{\scriptsize ONMIN}}[ns] \times f_{\mbox{\scriptsize OSC}}[\mbox{\scriptsize MHz}] \times 10^{-3}$ 

<sup>(\*2)</sup> Design reference value. This parameter is provided only for reference. (\*3) Current limit denotes the level of detection at peak of coil current.

<sup>(\*4)</sup> For the USP-6C Package only.

<sup>(\*5)</sup> EFFI = {(output voltage) x (output current)} / {(input voltage) x (input current)} x 100

# ■ELECTRICAL CHARACTERISTICS(Continued)

XC9267 series Ta=25°C

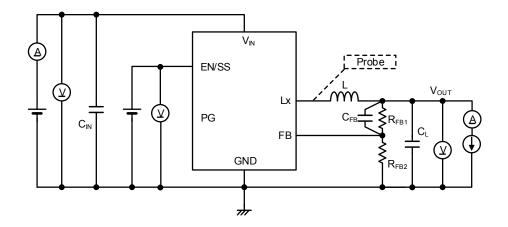
10 20 0							
ARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
FB "H" Current	Іғвн	V <sub>IN</sub> =V <sub>EN/SS</sub> =36V, V <sub>FB</sub> =3.0V	-0.1	-	0.1	μA	4
FB "L" Current	I <sub>FBL</sub>	V <sub>IN</sub> =V <sub>EN/SS</sub> =36V, V <sub>FB</sub> =0V	-0.1	1	0.1	μA	4
EN/SS "H" Voltage	Ven/ssh	V <sub>EN/SS</sub> =0.3V→2.5V, V <sub>FB</sub> =0.71V V <sub>EN/SS</sub> Voltage when Lx pin voltage changes from "L" level to "H" level	2.5	ı	36	٧	2
EN/SS "L" Voltage	V <sub>EN/SSL</sub>	V <sub>EN/SS</sub> =2.5V→0.3V, V <sub>FB</sub> =0.71V V <sub>EN/SS</sub> Voltage when Lx pin voltage changes from "H" level to "L" level	-	-	0.3	V	2
EN/SS "H" Current	I <sub>EN/SSH</sub>	V <sub>IN</sub> =V <sub>EN/SS</sub> =36V, V <sub>FB</sub> =0.825V	-	0.1	0.3	μA	4
EN/SS "L" Current	I <sub>EN/SSL</sub>	V <sub>IN</sub> =36V, V <sub>EN/SS</sub> =0V, V <sub>FB</sub> =0.825V	-0.1	1	0.1	μA	4
Thermal Shutdown Temperature	T <sub>TSD</sub>	Junction Temperature	-	150	-	°C	_
Hysteresis Width	Thys	Junction Temperature	-	25	-	°C	_

Test Condition: Unless otherwise stated, V<sub>IN</sub>=12V, V<sub>EN/SS</sub>=12V, V<sub>PG</sub>:OPEN (\*4)

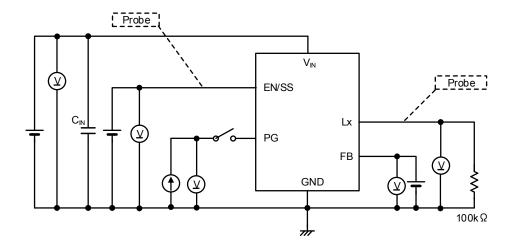
<sup>(\*4)</sup> For the USP-6C Package only.

# **■**TEST CIRCUITS

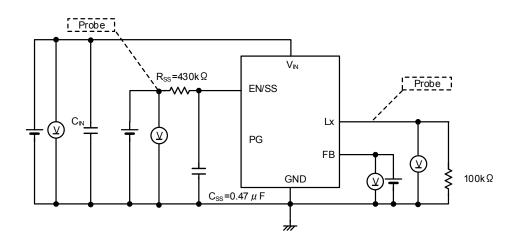
### CIRCUIT(1)



### CIRCUIT②



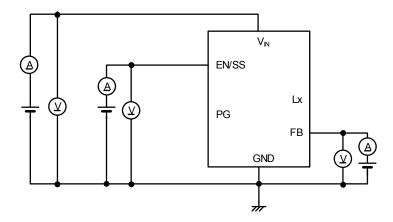
### CIRCUIT®



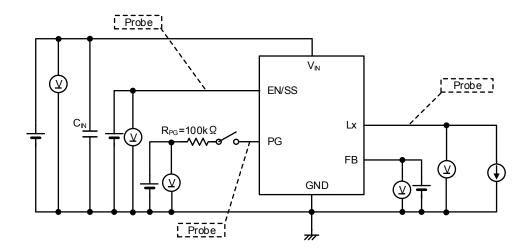
<sup>\*</sup> PG Pin is USP-6C Package only.

# ■TEST CIRCUITS(Continued)

## CIRCUIT4

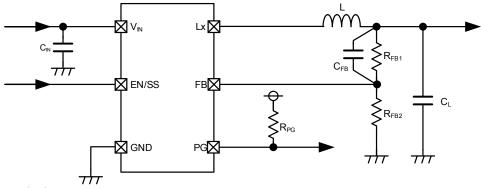


### CIRCUIT®



<sup>\*</sup> PG Pin is USP-6C Package only.

# **■TYPICAL APPLICATION CIRCUIT**



<Inductance value setting>

For the XC9267 Series, operation is optimized by setting the following inductance value according to the set frequency and setting output voltage.

Foscset: Frequency setting Voutset: Output voltage setting

### [Typical Examples]

	Foscset	conditions	MANUFACTURER	PRODUCT NUMBER	VALUE	
		4)/ (2)/	TDK	CLF5030NIT-3R3N	2.2.11	
		1V <v<sub>OUTSET≦2V</v<sub>	Coilcraft	XEL4030-332ME	3.3µH	
		01/41/	TDK	CLF5030NIT-4R7N	47.11	
	1.2MHz	2V <v<sub>OUTSET≦3.3V</v<sub>	Coilcraft	XEL4030-472ME	4.7μH	
		20141	TDK	CLF5030NIT-6R8N	0.0.11	
		3.3V <v<sub>OUTSET≦6V</v<sub>	Coilcraft	XEL4030-682ME	6.8µH	
		6V <voutset≦25v< td=""><td>TDK</td><td>CLF5030NIT-100N</td><td>10μH</td></voutset≦25v<>	TDK	CLF5030NIT-100N	10μH	
L		1V <voutset≤2v< td=""><td>TDK</td><td>CLF5030NIT-1R5N</td><td>1.5</td></voutset≤2v<>	TDK	CLF5030NIT-1R5N	1.5	
		IV < VOUTSET ≦2V	Coilcraft	XEL4030-152ME	- 1.5μH	
		2V <voutset≦3.3v< td=""><td>TDK</td><td>CLF5030NIT-2R2N</td><td>2 2</td></voutset≦3.3v<>	TDK	CLF5030NIT-2R2N	2 2	
	2.2MHz		Coilcraft	XEL4030-222ME	- 2.2μH	
	∠.∠IVI⊓∠	3.3V <voutset≦6v< td=""><td>TDK</td><td>CLF5030NIT-3R3N</td><td>3.3µH</td></voutset≦6v<>	TDK	CLF5030NIT-3R3N	3.3µH	
		3.3V < VOUTSET ≥ 0V	Coilcraft	XEL4030-332ME	3.3μΠ	
		6V <v<sub>OUTSET≦25V</v<sub>	TDK	CLF5030NIT-4R7N	4.7	
		0V < VOUTSET ≧23V	Coilcraft	XEL4030-472ME	- 4.7μH	
	1.2MHz	V <sub>IN</sub> <20V	TDK	C2012X6S1H475K125AC	4.7µF/50V	
C <sub>IN</sub>	1.ZIVITIZ	V <sub>IN</sub> ≧20V	TDK	C2012X6S1H475K125AC	4.7µF/50V 2parallel	
CIN	2.2MHz	V <sub>IN</sub> <20V	TDK	C2012X7R1H225K125AC	2.2µF/50V	
	Z.ZIVITZ	V <sub>IN</sub> ≧20V	TDK	TDK C2012X7R1H225K125AC		
				C2012X7R1A106K125AC	10µF/10V 2parallel	
CL	-	-	TDK	C3216X7R1E106K160AB	10µF/35V 2parallel	
				C3225X7R1H106M250AC	10µF/50V 2parallel	

# ■TYPICAL APPLICATION CIRCUIT(Continued)

#### < Output voltage setting >

The output voltage can be set by adding an external dividing resistor.

The output voltage is determined by the equation below based on the values of R<sub>FB1</sub> and R<sub>FB2</sub>.

$$V_{OUT}$$
=0.75 $V$ × ( $R_{FB1}$ + $R_{FB2}$ )/ $R_{FB2}$   
With  $R_{FB2}$  $\leq$ 400 $k\Omega$ 

#### <C<sub>FB</sub> setting>

Adjust the value of the phase compensation speed-up capacitor CFB using the equation below.

$$C_{FB} = \frac{1}{2\pi \times fz fb \times R_{FB1}}$$

A target value for fzfb of about  $fzfb = \frac{1}{2\pi\sqrt{C_L \times L}}$  is optimum.

#### [Setting Example]

To set output voltage to 5V with fosc=1.2MHz, C<sub>L</sub>=10µF×2, L=6.8µH

When R<sub>FB1</sub>=680k $\Omega$ , R<sub>FB2</sub>=120k $\Omega$ , V<sub>OUTSET</sub>=0.75V× (680k $\Omega$ +120k $\Omega$ ) / 120k $\Omega$ =5.0V

And fzfb is set to a target of 13.65 kHz using the above equation,

 $C_{FB}=1/(2\times\pi\times13.65 \text{ kHz}\times680\text{k}\Omega)=17.15\text{pF}$ . A capacitor of E24 series is 18pF.

	XC9267B75Cxx / f <sub>OSC</sub> =1.2MHz						
Voutset	R <sub>FB1</sub>	R <sub>FB2</sub>	L	Сғв	fzfb		
1.2V	120kΩ	200kΩ	3.3µH	68pF	19.6kHz		
3.3V	510kΩ	150kΩ	4.7µH	18pF	16.4kHz		
5.0V	680kΩ	120kΩ	6.8µH	18pF	13.7kHz		
12V	1500kΩ	100kΩ	10µH	9pF	11.3kHz		

	XC9267B75Dxx / f <sub>OSC</sub> =2.2MHz							
Voutset	R <sub>FB1</sub>	R <sub>FB2</sub>	L	Сғв	fzfb			
1.2V	120kΩ	200kΩ	1.5µH	47pF	29.1kHz			
3.3V	510kΩ	150kΩ	2.2µH	12pF	24.0kHz			
5.0V	680kΩ	120kΩ	3.3µH	12pF	19.6kHz			
12V	1500kΩ	100kΩ	4.7µH	6.2pF	16.4kHz			

#### <Soft-start Time Setting>

The soft-start time can be adjusted by adding a capacitor and a resistor to the EN/SS pin.

Soft-start time (tss2) is approximated by the equation below according to values of V<sub>EN/SS</sub>, R<sub>SS</sub>, and C<sub>SS</sub>.

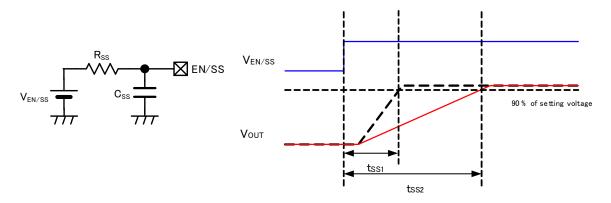
$$t_{ss2}=C_{ss}\times R_{ss}\times In (V_{EN/SS}/(V_{EN/SS}-1.45))$$

#### [Setting Example]

When  $C_{SS}$ =0.47 $\mu$ F,  $R_{SS}$ =430 $k\Omega$  and  $V_{EN/SS}$ =12V,  $t_{SS2}$ =0.47x10<sup>-6</sup> x 430 x 10<sup>3</sup> x (In (12/ (12-1.45)) =26ms (Approx.)

\*The soft-start time is the time from the start of V<sub>EN/SS</sub> until the output voltage reaches 90% of the set voltage.

If the EN/SS pin voltage rises steeply without connecting  $C_{SS}$  and  $R_{SS}$  ( $R_{SS}=0\Omega$ ), Output rises with taking the soft-start time of  $t_{SS1}=2.0$ ms (TYP.) which is fixed internally.

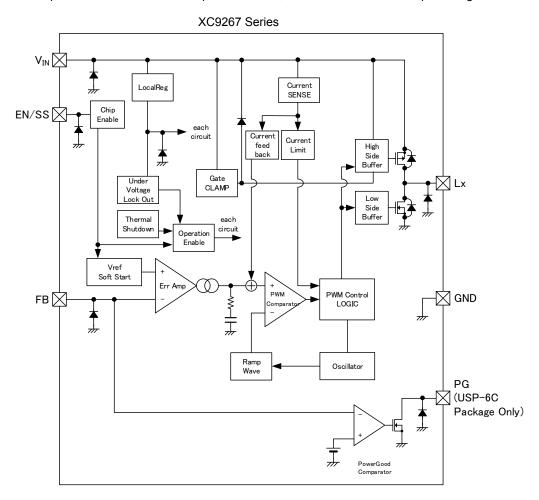


### ■OPERATIONAL EXPLANATION

The XC9267 series consists internally of a reference voltage supply with soft-start function, error amp, PWM comparator,ramp wave circuit, oscillator circuit, phase compensation (Current feedback) circuit, current limiting cir

The voltage feedback from the FB pin is compared to the internal reference voltage by the error amp, the output from the error amp is phase compensated, and the signal is input to the PWM comparator to determine the ON time of switching during PWM operation. The output signal from the error amp is compared to the ramp wave by the PWM comparator, and the output is sent to the buffer drive circuit and output from the LX pin as the duty width of switching. This operation is performed continuously to stabilize the output voltage.

The driver transistor current is monitored at each switching by the phase compensation (Current feedback) circuit, and the output signal from the error amp is modulated as a multi-feedback signal. This allows a stable feedback system to be obtained even when a low ESR capacitor such as a ceramic capacitor is used, and this stabilizes the output voltage.



\* Diodes inside the circuits are ESD protection diodes and parasitic diodes.

#### <Reference voltage source>

The reference voltage source provides the reference voltage to ensure stable output voltage of the DC/DC converter.

#### <Oscillator circuit>

The oscillator circuit determines switching frequency. 1.2MHz or 2.2MHz is available for the switching frequency. Clock pulses generated in this circuit are used to produce ramp waveforms needed for PWM operation.

#### <Error amplifier>

The error amplifier is designed to monitor output voltage. The amplifier compares the reference voltage with the feedback voltage divided by the internal voltage divider, R<sub>FB1</sub> and R<sub>FB2</sub>. When a voltage is lower than the reference voltage, then the voltage is fed back, the output voltage of the error amplifier increases. The error amplifier output is fixed internally to deliver an optimized signal to the mixer.

### ■ OPERATIONAL EXPLANATION(Continued)

#### <Current limiting>

The current limiting circuit of the XC9267 series monitors the current that flows through the High-side driver transistor and Low-side driver transistor, and when over-current is detected, the current limiting function activates.

#### (1) High-side driver Tr. current limiting

The current in the High-side driver Tr. is detected to equivalently monitor the peak value of the coil current. The High-side driver Tr. current limiting function forcibly turns off the High-side driver Tr. when the peak value of the coil current reaches the High-side driver current limit value I<sub>LIMH</sub>.

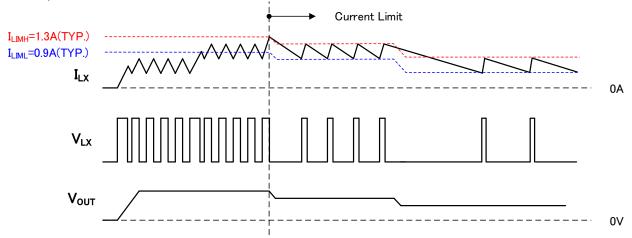
High-side driver Tr. current limit value ILIMH=1.3A (TYP.)

#### (2) Low-side driver Tr. current limiting

The current in the Low-side driver Tr. is detected to equivalently monitor the bottom value of the coil current. The Low-side driver Tr. current limiting function operates when the High-side driver Tr. current limiting value reaches I<sub>LIMH</sub>. The Low-side driver Tr. current limiting function prohibits the High-side driver Tr. from turning on in an over-current state where the bottom value of the coil current is higher than the Low-side driver Tr. current limit value I<sub>LIML</sub>.

Low side driver Tr. current limit value ILIML=0.9A (TYP.)

The current foldback circuit operates control to lower the switching frequency f<sub>OSC</sub>. When the over-current state is released, normal operation resumes.



#### <Soft-start function>

The output voltage of XC9267 rises with soft start by slowly raising the reference voltage. The rise time of this reference voltage is the soft start time. The soft-start time is set to 2.0ms (TYP.) which is fixed internally or to the time set by adding a capacitor and a resistor to the EN / SS pin whichever is later.

#### <Thermal shutdown>

The thermal shutdown (TSD) as an over temperature limit is built in the XC9267 series.

When the junction temperature reaches the detection temperature, the driver transistor is forcibly turned off. When the junction temperature falls to the release temperature while in the output stop state, restart takes place by soft-start.

#### <UVLO>

When the  $V_{\text{IN}}$  pin voltage falls below 2.7V (TYP.), the driver transistor is forcibly turned off to prevent false pulse output due to instable operation of the internal circuits. When the  $V_{\text{IN}}$  pin voltage rises above 2.8V (TYP.), the UVLO function is released, the soft-start function activates, and output start operation begins. Stopping by UVLO is not shutdown; only pulse output is stopped and the internal circuits continue to operate.

#### <Power good>

On USP-6C Package, the output state can be monitored using the power good function. When the FB voltage drops below 90% (TYP.), the PG pin outputs an "L" signal. The PG pin is an Nch open drain output, therefore a pull-up resistance (approx.  $100k\Omega$ ) must be connected to the PG pin.

# XC9267 Series

### ■NOTE ON USE

- 1) For the phenomenon of temporal and transitional voltage decrease or voltage increase, the IC may be damaged or deteriorated if IC is used beyond the absolute MAX. specifications.
- 2) Make sure that the absolute maximum ratings of the external components and of this IC are not exceeded.
- 3) The DC/DC converter characteristics depend greatly on the externally connected components as well as on the characteristics of this IC, so refer to the specifications and standard circuit examples of each component when carefully considering which components to select.

Be especially careful of the capacitor characteristics and use X7R or X5R (EIA standard) ceramic capacitors.

The capacitance decrease caused by the bias voltage may become remarkable depending on the external size of the capacitor.

4) The DC/DC converter of this IC uses a current-limiting circuit to monitor the coil peak current. If the potential dropout voltage is large or the load current is large, the peak current will increase, which makes it easier for current limitation to be applied which in turn could cause the operation to become unstable. When the peak current becomes large, adjust the coil inductance and sufficiently check the operation.

The following formula is used to show the peak current.

Peak Current: lpk = (V<sub>IN</sub> - V<sub>OUT</sub>) × V<sub>OUT</sub> / V<sub>IN</sub> / (2 × L × f<sub>OSC</sub>) + I<sub>OUT</sub>

L: Coil Inductance [H]

fosc: Oscillation Frequency [Hz]

IOUT: Load Current [A]

- 5) If there is a large dropout voltage, a circuit delay could create the ramp-up of coil current with staircase waveform exceeding the current limit.
- 6) Even in the PWM control, the intermittent operation occurs and the ripple voltage becomes higher, when the minimum On Time is faster than 85ns (typ.) as well as the dropout voltage is large and output current is small.
- 7) The ripple voltage could be increased when switching from discontinuous conduction mode to continuous conduction mode and at switching to 100% Duty cycle. Please evaluate IC well on customer's PCB.
- 8) If the voltage at the EN/SS Pin does not start from 0V but it is at the midpoint potential when the power is switched on, the soft start function may not work properly and it may cause the larger inrush current and bigger ripple voltages.
- 9) Torex places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using Torex products in their systems.

# ■NOTE ON USE(Continued)

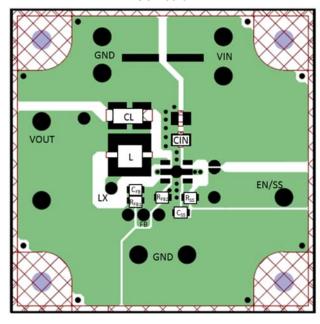
10) Instructions of pattern layouts

The operation may become unstable due to noise and/or phase lag from the output current when the wire impedance is high, please place the input capacitor( $C_{IN}$ ) and the output capacitor ( $C_{L}$ ) as close to the IC as possible.

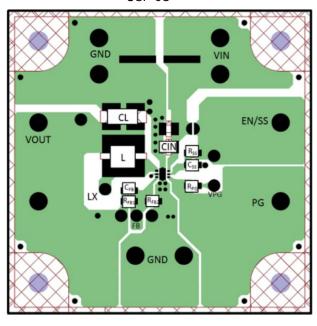
- (1) In order to stabilize  $V_{IN}$  voltage level, we recommend that a by-pass capacitor ( $C_{IN}$ ) be connected as close as possible to the  $V_{IN}$  and GND pins.
- (2) Please mount each external component as close to the IC as possible.
- (3) Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- (4) Make sure that the GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- (5) Please note that internal driver transistors bring on heat because of the load current and ON resistance of Highside driver transistor, Lowside driver transistor. Please make sure that the heat is dissipated properly, especially at higher temperatures.

<Reference Pattern Layout>

SOT-89-5



USP-6C

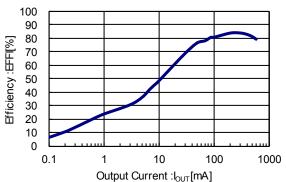


### ■TYPICAL PERFORMANCE CHARACTERISTICS

(1) Efficiency vs. Output Current

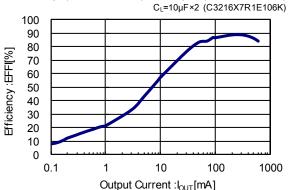
XC9267B75Cxx  $(V_{IN}=12V, V_{OUT}=3.3V, f_{OSC}=1.2MHz)$ 

L=4.7  $\mu$  H(CLF5030NIT-4R7), C<sub>IN</sub>=4.7  $\mu$  F×2(C2012X6S1H475K) C<sub>L</sub>=10  $\mu$  F×2 (C3216X7R1E106K)



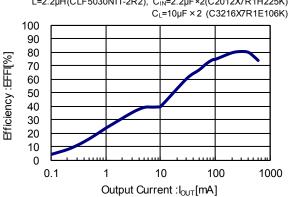
### XC9267B75Cxx $(V_{IN}=12V, V_{OUT}=5V, f_{OSC}=1.2MHz)$

L=6.8 $\mu$ H(CLF5030NIT-6R8), C<sub>IN</sub>=4.7 $\mu$ F×2(C2012X6S1H475K) C<sub>L</sub>=10µF×2 (C3216X7R1E106K)



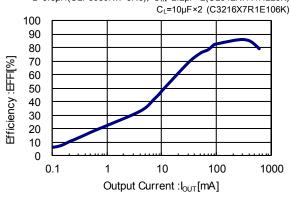
XC9267B75Dxx  $(V_{IN}=12V, V_{OUT}=3.3V, f_{OSC}=2.2MHz)$ 

L=2.2 $\mu$ H(CLF5030NIT-2R2), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K)



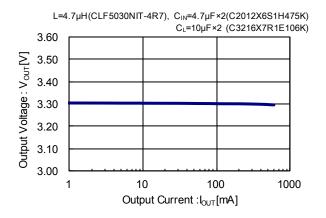
XC9267B75Dxx  $(V_{IN}=12V, V_{OUT}=5V, f_{OSC}=2.2MHz)$ 

L=3.3µH(CLF5030NIT-3R3), C<sub>IN</sub>=2.2µF×2(C2012X7R1H225K)

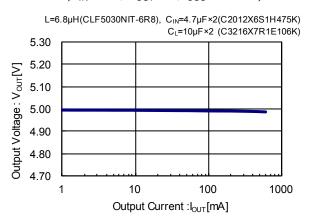


#### (2) Output Voltage vs. Output Current

XC9267B75Cxx  $(V_{IN}=12V, V_{OUT}=3.3V, f_{OSC}=1.2MHz)$ 



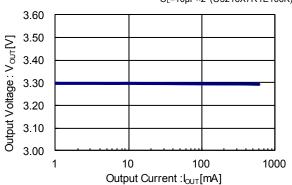
XC9267B75Cxx  $(V_{IN}=12V, V_{OUT}=5V, f_{OSC}=1.2MHz)$ 



(2) Output Voltage vs. Output Current

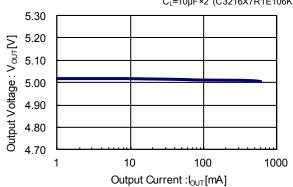
 $\begin{array}{c} XC9267B75Dxx\\ (V_{IN}\text{=}12V,\,V_{OUT}\text{=}3.3V,\,f_{OSC}\text{=}2.2MHz) \end{array}$ 

L=2.2 $\mu$ H(CLF5030NIT-2R2), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)



XC9267B75Dxx( $V_{IN}$ =12V,  $V_{OUT}$ =5V,  $f_{OSC}$ =2.2MHz)

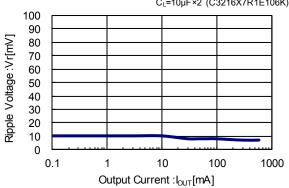
L=3.3 $\mu$ H(CLF5030NIT-3R3), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)



(3) Ripple Voltage vs. Output Current

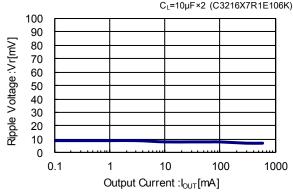
XC9267B75Cxx (V<sub>IN</sub>=12V, V<sub>OUT</sub>=5V, f<sub>OSC</sub>=1.2MHz)

 $\begin{array}{lll} L{=}6.8\mu H (CLF5030NIT{-}6R8), & C_{IN}{=}4.7\mu F{\times}2 (C2012X6S1H475K) \\ & C_{L}{=}10\mu F{\times}2 & (C3216X7R1E106K) \end{array}$ 



XC9267B75Dxx (V<sub>IN</sub>=12V, V<sub>OUT</sub>=5V, f<sub>OSC</sub>=2.2MHz)

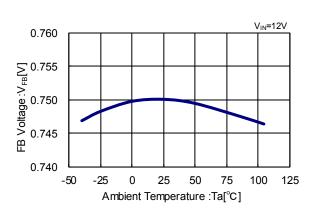
 $\begin{array}{lll} L{=}3.3\mu H (CLF5030NIT{-}3R3), & C_{IN}{=}2.2\mu F{\times}2 (C2012X7R1H225K) \\ & C_{L}{=}10\mu F{\times}2 & (C3216X7R1E106K) \end{array}$ 



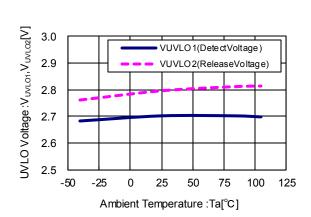
(4) FB Voltage vs. Ambient Temperature

(5) UVLO Voltage vs. Ambient Temperature

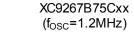


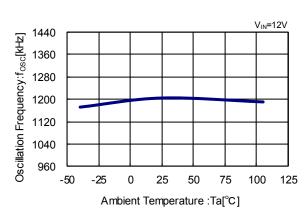


#### XC9267B75xxx

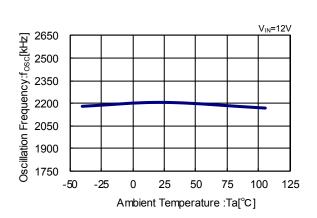


(6) Oscillation Frequency vs. Ambient Temperature



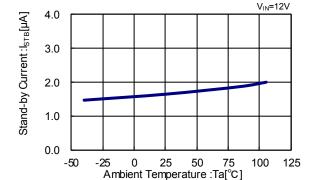


### XC9267B75Dxx $(f_{OSC}=2.2MHz)$



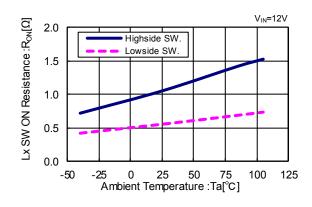
(7) Stand-by Current vs. Ambient Temperature

#### XC9267B75xxx



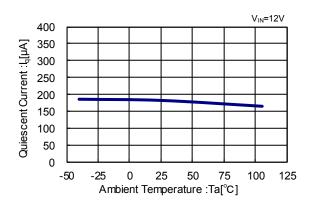
(8) Lx SW ON Resistance vs. Ambient Temperature

#### XC9267B75xxx

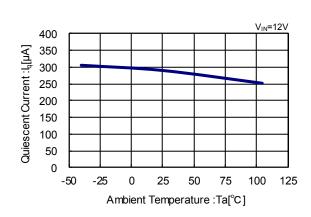


(9) Quiescent Current vs. Ambient Temperature

XC9267B75Cxx( $f_{OSC}=1.2MHz$ )

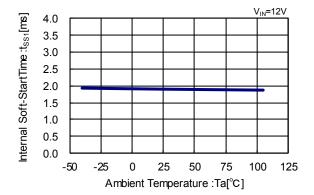


XC9267B75Dxx( $f_{OSC}=2.2MHz$ )

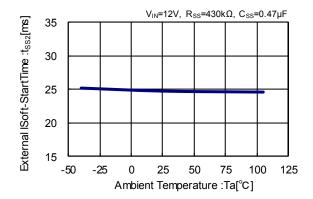


(10) Internal Soft-Start Time vs. Ambient Temperature (11) External Soft-Start Time vs. Ambient Temperature



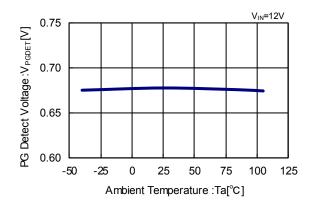


XC9267B75xxx



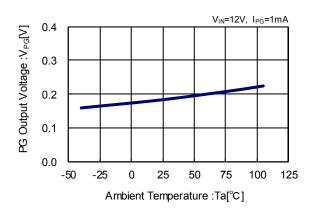
(12) PG Detect Voltage vs. Ambient Temperature

XC9267B75xxx



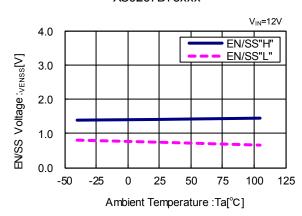
(13) PG Output Voltage vs. Ambient Temperature

#### XC9267B75xxx



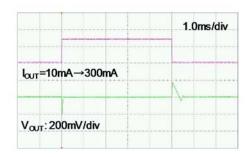
(14) EN/SS Voltage vs. Ambient Temperature

#### XC9267B75xxx

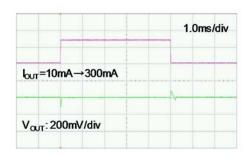


(15) Load Transient Response

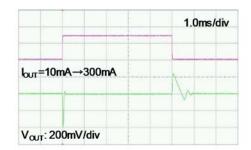
 $\begin{array}{c} XC9267B75Cxx, \ f_{OSC} = 1.2MHz \\ V_{IN} = 12V, \ V_{OUT} = 3.3V, \ I_{OUT} = 10mA \rightarrow 300mA \\ L = 4.7 \mu H (CLF5030NIT-4R7), \ C_{IN} = 4.7 \mu F \times 2 (C2012X6S1H475K) \\ C_{L} = 10 \mu F \times 2 \ (C3216X7R1E106K) \end{array}$ 



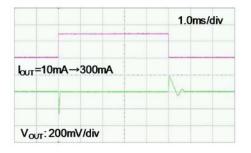
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =24V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =10mA $\rightarrow$ 300mA L=4.7μH(CLF5030NIT-4R7),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



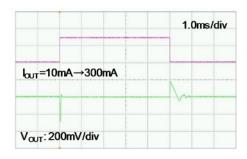
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =12V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =10mA $\rightarrow$ 300mA L=6.8μH(CLF5030NIT-6R8),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =24V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =10mA $\rightarrow$ 300mA L=6.8μH(CLF5030NIT-6R8),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)

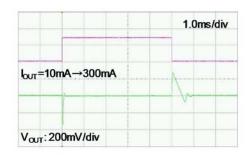


 $\begin{array}{c} \text{XC9267B75Dxx,} \ \ f_{OSC} = 2.2 \text{MHz} \\ \text{$V_{\text{IN}}$=} 12 \text{V,} \ \ V_{\text{OUT}} = 3.3 \text{V,} \ \ I_{\text{OUT}} = 10 \text{mA} \rightarrow 300 \text{mA} \\ \text{$L$=} 2.2 \mu \text{H(CLF5030NIT-2R2),} \ \ \ C_{\text{IN}} = 2.2 \mu \text{F} \times 2 \text{(C2012X7R1H225K)} \\ \text{$C_{\text{L}}$=} 10 \mu \text{F} \times 2 \text{ (C3216X7R1E106K)} \\ \end{array}$ 

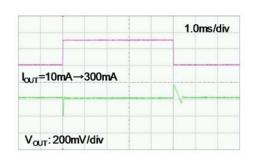


(15) Load Transient Response

XC9267B75Dxx、 $f_{OSC}$ =2.2MHz  $V_{IN}$ =12V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =10mA→300mA L=3.3 $\mu$ H(CLF5030NIT-3R3),  $C_{IN}$ =2.2 $\mu$ F×2(C2012X7R1H225K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)

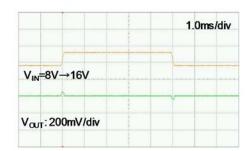


XC9267B75Dxx、 $f_{OSC}$ =2.2MHz V<sub>IN</sub>=24V, V<sub>OUT</sub>=5.0V,  $I_{OUT}$ =10mA $\rightarrow$ 300mA L=3.3μH(CLF5030NIT-3R3), C<sub>IN</sub>=2.2μF×2(C2012X7R1H225K) C<sub>L</sub>=10μF×2 (C3216X7R1E106K)

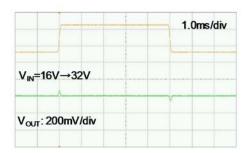


(16) Input Transient Response

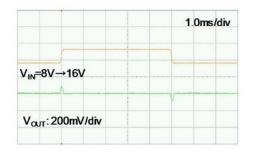
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =8V $\rightarrow$ 16V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA L=4.7μH(CLF5030NIT-4R7),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



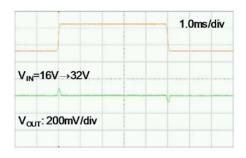
XC9267B75Cxx,  $f_{OSC}$ =1.2MHz  $V_{IN}$ =16V $\rightarrow$ 32V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA L=4.7μH(CLF5030NIT-4R7),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =8V $\rightarrow$ 16V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =300mA L=6.8μH(CLF5030NIT-6R8),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



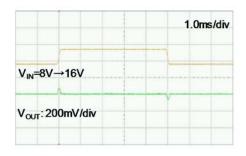
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =16V $\rightarrow$ 32V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =300mA L=6.8μH(CLF5030NIT-6R8),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



(16) Input Transient Response

XC9267B75Dxx,  $f_{OSC}$ =2.2MHz

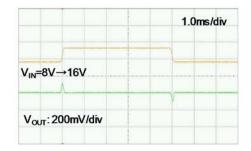
L=2.2 $\mu$ H(CLF5030NIT-2R2), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)



XC9267B75Dxx、f<sub>osc</sub>=2.2MHz

 $V_{IN}$ =8V $\rightarrow$ 16V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =300mA

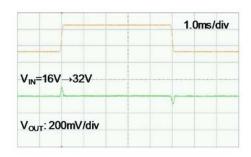
L=3.3 $\mu$ H(CLF5030NIT-3R3), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)



XC9267B75Dxx、fosc=2.2MHz

 $V_{IN}$ =16V $\rightarrow$ 32V,  $V_{OUT}$ =5.0V,  $I_{OUT}$ =300mA

L=3.3 $\mu$ H(CLF5030NIT-3R3), C<sub>IN</sub>=2.2 $\mu$ F×2(C2012X7R1H225K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)

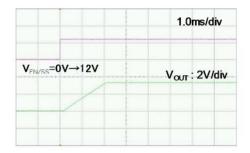


(17) EN/SS Rising Response

XC9267B75Cxx,  $f_{OSC}=1.2MHz$ 

 $V_{IN}$ =12V,  $V_{ENSS}$ =0 $\rightarrow$ 12V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA

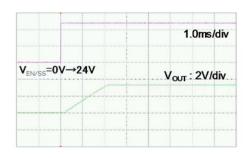
L=4.7 $\mu$ H(CLF5030NIT-4R7), C<sub>IN</sub>=4.7 $\mu$ F×2(C2012X6S1H475K) C<sub>L</sub>=10 $\mu$ F×2 (C3216X7R1E106K)



XC9267B75Cxx,  $f_{OSC}=1.2MHz$ 

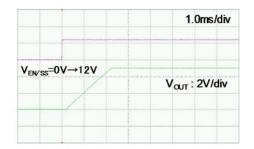
 $V_{IN}$ =24V,  $V_{ENSS}$ =0 $\rightarrow$ 24V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA

 $\begin{array}{c} L{=}4.7\mu H (CLF5030NIT{-}4R7), \ \ C_{IN}{=}4.7\mu F{\times}2 (C2012X6S1H475K) \\ C_{L}{=}10\mu F{\times}2 \ \ (C3216X7R1E106K) \end{array}$ 

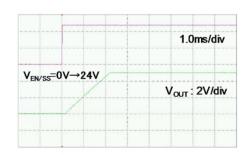


(17) EN/SS Rising Response

XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =12V,  $V_{ENSS}$ =0→12V,  $V_{OUT}$ =5V,  $I_{OUT}$ =300mA L=6.8 $\mu$ H(CLF5030NIT-6R8),  $C_{IN}$ =4.7 $\mu$ F×2(C2012X6S1H475K)  $C_{IL}$ =10 $\mu$ F×2 (C3216X7R1E106K)



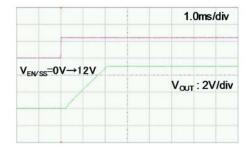
$$\begin{split} & XC9267B75Cxx, \ f_{OSC} = 1.2MHz \\ & V_{IN} = 24V, \ V_{ENSS} = 0 \rightarrow 24V, \ V_{OUT} = 5V, \ I_{OUT} = 300mA \\ & L = 6.8 \mu H (CLF5030NIT-6R8), \ C_{IN} = 4.7 \mu F \times 2 (C2012X6S1H475K) \\ & C_{L} = 10 \mu F \times 2 \ (C3216X7R1E106K) \end{split}$$



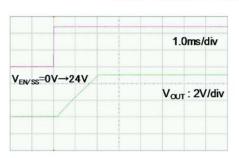
XC9267B75Dxx、 $f_{OSC}$ =2.2MHz  $V_{IN}$ =12V,  $V_{ENSS}$ =0→12V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA L=2.2 $\mu$ H(CLF5030NIT-2R2),  $C_{IN}$ =2.2 $\mu$ F×2(C2012X7R1H225K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)



XC9267B75Dxx、 $f_{OSC}$ =2.2MHz  $V_{IN}$ =12V,  $V_{ENSS}$ =0→12V,  $V_{OUT}$ =5V,  $I_{OUT}$ =300mA L=3.3 $\mu$ H(CLF5030NIT-3R3),  $C_{IN}$ =2.2 $\mu$ F×2(C2012X7R1H225K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)

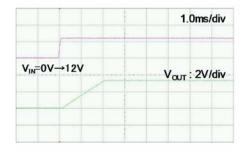


XC9267B75Dxx、 $f_{OSC}$ =2.2MHz  $V_{IN}$ =24V,  $V_{ENSS}$ =0→24V,  $V_{OUT}$ =5V,  $I_{OUT}$ =300mA L=3.3 $\mu$ H(CLF5030NIT-3R3),  $C_{IN}$ =2.2 $\mu$ F×2(C2012X7R1H225K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)

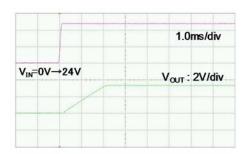


(18) VIN Rising Response

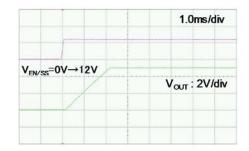
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =0 $\rightarrow$ 12V,  $V_{ENSS}$ =0 $\rightarrow$ 12V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA L=4.7μH(CLF5030NIT-4R7),  $C_{IN}$ =4.7μF×2(C2012X6S1H475K)  $C_{L}$ =10μF×2 (C3216X7R1E106K)



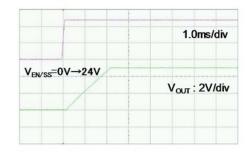
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =0 $\rightarrow$ 24V,  $V_{ENSS}$ =0 $\rightarrow$ 24V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =300mA L=4.7 $\mu$ H(CLF5030NIT-4R7),  $C_{IN}$ =4.7 $\mu$ F×2(C2012X6S1H475K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)



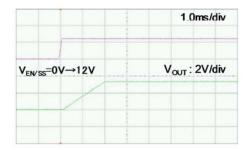
XC9267B75Cxx、 $f_{OSC}$ =1.2MHz  $V_{IN}$ =0 $\rightarrow$ 12V,  $V_{ENSS}$ =0 $\rightarrow$ 12V,  $V_{OUT}$ =5V,  $I_{OUT}$ =300mA L=6.8 $\mu$ H(CLF5030NIT-6R8),  $C_{IN}$ =4.7 $\mu$ F×2(C2012X6S1H475K)  $C_{L}$ =10 $\mu$ F×2 (C3216X7R1E106K)



 $\begin{array}{c} XC9267B75Cxx, \ \ f_{OSC} = 1.2MHz \\ V_{IN} = 0 \rightarrow 24V, \ V_{ENSS} = 0 \rightarrow 24V, \ V_{OUT} = 5V, \ I_{OUT} = 300mA \\ L = 6.8 \mu H (CLF5030NIT-6R8), \ C_{IN} = 4.7 \mu F \times 2 (C2012X6S1H475K) \\ C_{L} = 10 \mu F \times 2 \ (C3216X7R1E106K) \end{array}$ 

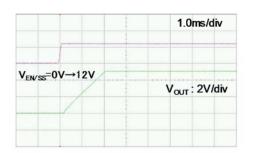


 $\begin{array}{c} \text{XC9267B75Dxx,} \ f_{\text{OSC}} = 2.2 \text{MHz} \\ \text{$V_{\text{IN}}$=0$$$\to$12V,} \ V_{\text{ENSS}} = 0$\to$12V,} \ V_{\text{OUT}} = 3.3 \text{V,} \ I_{\text{OUT}} = 300 \text{mA} \\ \text{$L$=2.2$$$\mu$H(CLF5030NIT-2R2),} \ C_{\text{IN}} = 2.2 \text{$\mu$F}$$\times$2 (C2012X7R1H225K)} \\ C_{\text{L}} = 10 \text{$\mu$F}$\times$2 (C3216X7R1E106K)} \end{array}$ 

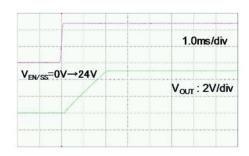


(18) VIN Rising Response

 $\begin{array}{c} \text{XC9267x75D\_} \text{ } f_{OSC} = 2.2 \text{MHz} \\ \text{$V_{\text{IN}}$=0$\rightarrow$12V, $V_{\text{ENSS}}$=0$\rightarrow$12V, $V_{\text{OUT}}$=5V, $I_{\text{OUT}}$=300 mA} \\ \text{$L$=3.3$\mu$H(CLF5030NIT-3R3N-D), $C_{\text{IN}}$=2.2$\mu$F$\times$2(C2012X7R1H225K)} \\ \text{$C_{\text{L}}$=10$\mu$F$\times$2 (C3216X7R1E106K)} \end{array}$ 

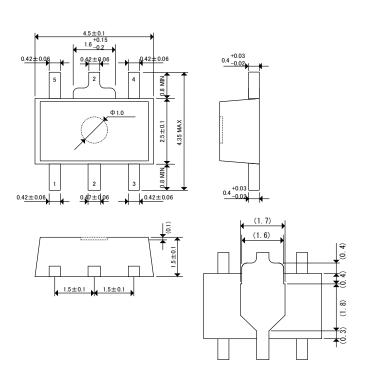


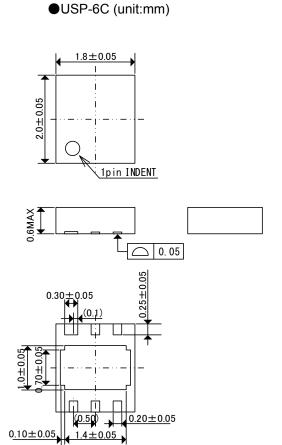
$$\begin{split} & XC9267B75Dxx, \ f_{OSC} = & 2.2MHz \\ & V_{\text{IN}} = & 0 \rightarrow 24V, \ V_{\text{ENSS}} = & 0 \rightarrow 24V, \ V_{\text{OUT}} = & 5V, \ I_{\text{OUT}} = & 3.00mA \\ & L = & 3.3 \mu \text{H(CLF5030NIT-3R3N-D)}, \ C_{\text{IN}} = & 2.2 \mu \text{F} \times 2 \text{(C2012X7R1H225K)} \\ & C_{\text{L}} = & 10 \mu \text{F} \times 2 \text{ (C3216X7R1E106K)} \end{split}$$



# **■**PACKAGING INFORMATION

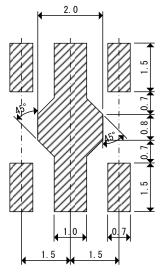
●SOT-89-5(unit:mm)





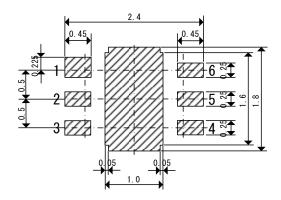
# ■ PACKAGING INFORMATION (Continued)

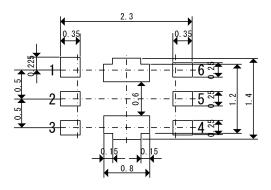
●SOT-89-5 Reference Pattern Layout (unit: mm)



●USP-6C Reference Pattern Layout (unit: mm)

●USP-6C Reference Metal Mask Design (unit: mm)





#### ●SOT-89-5 Power Dissipation (JEDEC)

Power dissipation data for the SOT-89-5 is shown in this page.

The value of power dissipation varies with the mount board conditions.

Please use this data as one of reference data taken in the described condition.

#### 1. Measurement Condition (Reference data)

Condition: Mount on a board Ambient: Natural convection Soldering: Lead (Pb) free

Board: The board using 4 copper layer.

(76.2mm×114.3mm···Area: about 8700mm²)

1st layer: No copper foil (Signal layer)

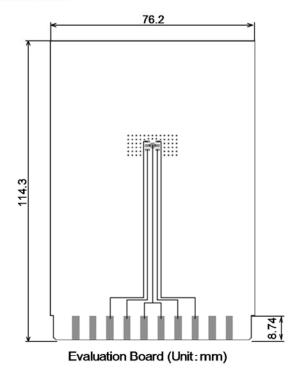
2nd layer: 70mm×70mm\_Connected to heat-sink.
3rd layer: 70mm×70mm\_Connected to heat-sink.

4th layer: No copper foil (Signal layer)

Material: Glass Epoxy (FR-4)

Thickness: 1.6mm

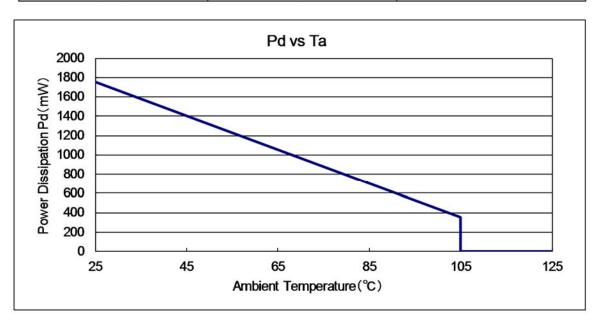
Through-hole:  $\phi$ 0.2mm x 60pcs



### 2. Power Dissipation vs. Ambient Temperature

Board Mount (Ti max = 125°C)

Ambient Temperature (°C)	Power Dissipation Pd (mW)	Thermal Resistance (°C/W)	
25	1750	E7 14	
105	350	57.14	



#### ●USP-6C(Insulation Paste/DAF) Power Dissipation (JEDEC)

Power dissipation data for the USP-6C(Insulation Paste/DAF) is shown in this page.

The value of power dissipation varies with the mount board conditions.

Please use this data as one of reference data taken in the described condition.

#### 1. Measurement Condition (Reference data)

Condition: Mount on a board Ambient: Natural convection Soldering: Lead (Pb) free

Board: The board using 4 copper layer.

(76.2mm×114.3mm···Area: about 8700mm<sup>2</sup>)

1st layer: No copper foil (Signal layer)

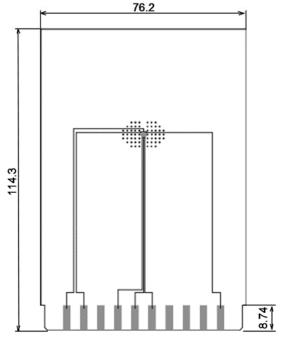
2nd layer: 70mm×70mm\_Connected to heat-sink.
3rd layer: 70mm×70mm\_Connected to heat-sink.

4th layer: No copper foil (Signal layer)

Material: Glass Epoxy(FR-4)

Thickness: 1.6mm

Through-hole: φ0.2mm x 60pcs

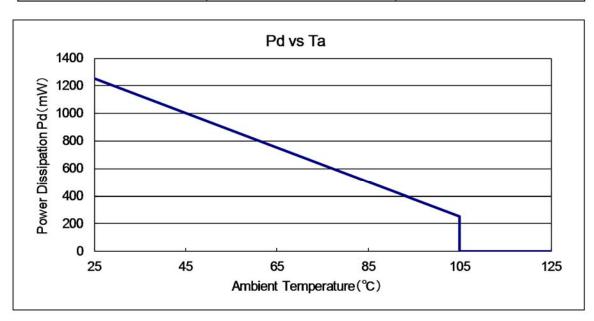


### Evaluation Board (Unit:mm)

### 2. Power Dissipation vs. Ambient Temperature

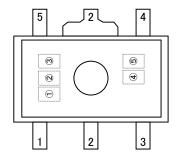
#### Board Mount (Ti max = 125°C)

Ambient Temperature (°C)	Power Dissipation Pd(mW)	Thermal Resistance (°C/W)	
25	1250	80.00	
105	250	80.00	



### **■**MARKING RULE

### ●SOT-89-5

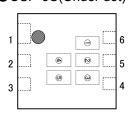


①② represents product series, products type,

MARK		DDODUCT CEDIES
1	2	PRODUCT SERIES
5	1	XC9267B75***-G

**XUSP-6C** Under dot

## ●USP-6C(Under dot)



3 represents Oscillation Frequency

MARK	Oscillation Frequency	PRODUCT SERIES
N	1.2MHz	XC9267B75C**-G
U	2.2MHz	XC9267B75D**-G

(4)⑤ represents production lot number 01~09, 0A~0Z, 11~9Z, A1~A9, AA~AZ, B1~ZZ repeated (G, I, J, O, Q, W excluded)\* No character inversion used.

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